# **LCR Parallel Resonance**

## **Expression for Resonant Frequency:**

Parallel resonance circuit is shown in fig(1). One branch of the resonant circuit consists resistance R in series with an inductance L. Capacitor C is connected parallel to the RL combination in other branch. An AC source is applied across this parallel combination. The admittance of the circuit is given by



or

$$
2\pi f_0 = \sqrt{\frac{1}{LC} - \frac{R^2}{L^2}} \qquad (\because \omega_0 = 2\pi f_0)
$$

2

 $f_0 = \frac{1}{2\pi} \sqrt{\frac{1}{LC} - \frac{R^2}{L^2}}$  $2\pi$  VLC L<sup>2</sup>  $=\frac{1}{2}$  $\sqrt{\frac{1}{2}$ π

or

where  $f_0$  = resonant frequency of LCR parallel circuit

#### **Dynamic Resistance**

The resistance at resonance is known as dynamic resistance. Consider the following equation.

--------------- (3)

at resonance  
\n
$$
\frac{1}{Z} = \frac{R}{(R^2 + \omega^2 L^2)} - j\omega \left[ \frac{L}{(R^2 + \omega^2 L^2)} - C \right]
$$
\nat resonance  
\n
$$
\left[ \frac{L}{(R^2 + \omega^2 L^2)} - C \right] = 0
$$
\n
$$
\therefore \frac{1}{Z} = \frac{R}{(R^2 + \omega^2 L^2)}
$$
\n
$$
\therefore \frac{1}{Z} = \frac{R}{(R^2 + \omega^2 L^2)} \quad \text{3.}\n\text{Substituting in Eqn(1), we get}
$$
\n
$$
\frac{1}{Z} = \frac{R}{L}
$$
\n
$$
\text{Or}
$$
\n
$$
\frac{1}{Z} = \frac{RC}{L}
$$
\n
$$
Z_0 = r_d = \frac{L}{RC}
$$
\n
$$
\text{1.}\n\text{or}
$$
\n
$$
Z_0 = r_d = \frac{L}{RC}
$$
\n
$$
\text{1.}\n\text{or}
$$
\n
$$
Z_0 = r_d = \frac{L}{RC}
$$
\n
$$
\text{2.}\n\text{1.
$$

at resonance

where  $r_d$  is known as dynamic resistance.

#### **Bandwidth**

Bandwidth is defined as the band of frequencies between two points on either side of resonant frequency where impedance falls to  $\frac{1}{\sqrt{2}}$ 2 times of value at resonance.

Impedance of the LCR parallel circuit is given by

$$
Z = \frac{(R + j\omega L)\frac{1}{j\omega C}}{(R + j\omega L + \frac{1}{j\omega C})}
$$



or

$$
Z = \frac{(R + j\omega L)}{j\omega C\left(R + j\omega L + \frac{1}{j\omega C}\right)}
$$

$$
Z = \frac{(R + j\omega L)}{(1 - j\omega L)^2}
$$

 $(j\omega RC - \omega^2 LC + 1)$ 

jωRC – ω<sup>2</sup>LC + 1)

 $\omega RC - \omega^2 LC +$ 

or

where  $R \ll \omega L$  then R can be neglected from the above equation.

$$
Z = \frac{j\omega L}{j\omega RC + (1 - \omega^{2} LC)}
$$

Z

Taking modulus on bothsides of the above equation, we get

$$
|Z| = \frac{\omega L}{\sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2}}
$$
 (1)

Multiplying numerator and denomination of eqn (1) with  $\frac{1}{10}$ ωRC , we get

$$
|Z| = \frac{\omega L x \frac{1}{\omega RC}}{\frac{1}{\omega RC} \sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2}}
$$
  
L

or

$$
|Z| = \frac{\overline{RC}}{\frac{1}{\omega RC} \sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2}}
$$

or 
$$
|Z| = \frac{Z_0}{\frac{1}{\omega RC} \sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2}} \qquad (\because Z_0 = \frac{L}{RC})
$$
  
or 
$$
|Z| = \frac{Z_0 \omega RC}{\sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2}}
$$
  
but 
$$
|Z| = \frac{Z_0}{\sqrt{C}}
$$

$$
\frac{Z_0}{\sqrt{2}} = \frac{Z_0 \text{ or } C}{\sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2}}
$$

$$
\frac{1}{\sqrt{2}} = \frac{\text{ or } C}{\sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2}}
$$

or

or 
$$
\sqrt{\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2} = \omega RC \sqrt{2}
$$
 (2)

squaring on both sides of equ(2), we get

$$
\omega^2 R^2 C^2 + (1 - \omega^2 LC)^2 = 2\omega^2 R^2 C^2
$$

or 
$$
(1 - \omega^2 LC)^2 = 2\omega^2 R^2 C^2 - \omega^2 R^2 C^2
$$

or 
$$
(1 - \omega^2 LC)^2 = \omega^2 R^2 C^2
$$

taking square root on both sides of the above equation, we get

$$
1 - \omega^2 LC = \pm \omega RC
$$
  
or  

$$
\omega^2 LC \pm \omega RC - 1 = 0
$$
 3

Eqn (3) is a quadratic equation in  $\omega$  and will yield two values. The two values  $\omega_1$  and  $\omega_2$  and they can be obtained as follows.

$$
= \frac{\pm RC \pm \sqrt{R^2C^2 + 4LC}}{2LC}
$$

$$
= \frac{\pm RC}{2LC} \pm \frac{\sqrt{R^2C^2 + 4LC}}{2LC}
$$

2 2  $R \mid \mathbb{R}^2 \mid 1$ 2L  $\sqrt{4L^2}$  LC  $=\frac{\pm R}{2L} \pm \sqrt{\left(\frac{R^2}{4L^2} + \frac{1}{LC}\right)}$ 

or

or

$$
= \frac{\pm RC}{2LC} \pm \sqrt{\left(\frac{R^2C^2}{4L^2C^2} + \frac{4LC}{4L^2C^2}\right)}
$$

or

as  $R \ll \omega L$ , therefore 2 2 R 4L can be ignored from the above equation, we get

R 1 2L LC ± = ± -------------------- (4)

Let  $\frac{R}{\sigma}$  $\frac{R}{2L} = \alpha$ , and we know that  $\sqrt{\frac{1}{LC}} = \omega_0$ LC  $=\omega_0$ , substituting these values in the equ (4) we get

$$
= \pm\,\alpha\pm\omega^{\vphantom{2}}_0
$$

Two roots can be written as follows

 $\omega_2 = \omega_0 + \alpha$  and  $\omega_1 = \omega_0 - \alpha$  (since angular frequency can not be negative)

Therefore Bandwidth ∴  $\Delta \omega = \omega_2 - \omega_1$ 

substituting  $\omega_1$  and  $\omega_2$  values in the above equation, we get

$$
\Delta \omega = \omega_0 + \alpha - (\omega_0 - \alpha)
$$
  
or  $\Delta \omega = \omega_0 + \alpha - \omega_0 + \alpha$   
or  $\Delta \omega = 2\alpha$  (5)  
but  $\alpha = \frac{R}{2L}$   
therefore  $\Delta \omega = \frac{2R}{2L}$   
or  $\Delta \omega = \frac{R}{L}$   
or  $\Delta \omega = \omega_2 - \omega_1 = (2\pi f_2 - 2\pi f_1) = \frac{R}{L}$   
or  $2\pi (f_2 - f_1) = \frac{R}{L}$   
 $(f_2 - f_1) = \frac{R}{2\pi L}$   
Therefore bandwidth  $\Delta f = (f_2 - f_1) = \frac{R}{2\pi L}$  (6)

### **Quality Factor ( Q – Factor)**

When voltage V is applied to the parallel LCR circuit, the current flowing through the circuit at resonance is given by

$$
I = \frac{V}{r_d} \qquad \qquad \text{---} \qquad (1)
$$

Where  $r_d$  is dynamic resistance =  $\frac{L}{R}$ RC

Substituting  $r_d$  value in eqn (1), we get

$$
I = \frac{V}{L/RC}
$$

$$
or \tI = \frac{VRC}{L}
$$

from above equation  $V = \frac{IL}{R}$ RC  $=\frac{12}{100}$  ----------- (2)

Voltage across the capacitor  $= V<sub>C</sub>$ 

Voltage across the inductor  $= V<sub>L</sub>$ 

From fig (4) voltage across capacitor  $(V_C)$  = voltage across inductor  $(V_L)$  = applied voltage (V) (Since the capacitor and the inductor are parallel to the applied voltage)

Voltage across capacitor

$$
V = I_C x \frac{1}{\omega_0 C}
$$

 $I_c = V \omega_0 C$  ---------- (3)

Substituting V value from eqn(2) in eqn(3) we get

$$
I_C = \frac{IL}{RC} \omega_0 C
$$
  
or  

$$
I_C = \frac{IL \omega_0}{R}
$$

Voltage across the inductor is  $V = I_L x \omega_0 L$ 

or 
$$
I_{L} = \frac{V}{\omega_{0} L}
$$
 -------(4)

substituting V value from eqn $(2)$  in eqn $(4)$ , we get

$$
I_{L} = \frac{IL}{RC} x \frac{1}{\omega_{0}L}
$$
  
or 
$$
I_{L} = \frac{I}{RC\omega_{0}} \quad \dots \dots \dots \tag{5}
$$

The Q-factor is defined as the magnification of the current at resonance

 $Q = \frac{I_C}{I} = \frac{I_L}{I}$ I I  $=\frac{1}{1}C =$ From eqn(4) we have I R  $=\frac{0}{x}$  $\therefore \qquad \frac{I_c}{I} = Q = \frac{\omega_0 L}{R}$ I R  $= Q = \frac{\omega_0 L}{R}$  ------------- (6)





From eqn (5) we have  $\frac{I_L}{I}$ 

From eqn (5) we have 
$$
\frac{I_{L}}{I} = \frac{1}{\omega_{0}RC}
$$

$$
\therefore \qquad \frac{I_{L}}{I} = Q = \frac{1}{\omega_{0}RC} \qquad \qquad \text{---} \qquad (7)
$$